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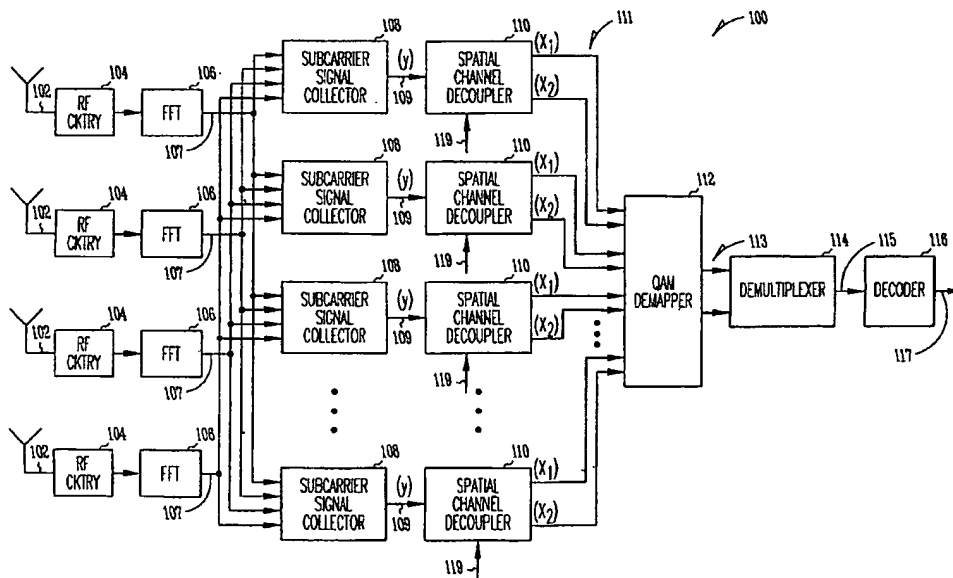
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(54) Title: MULTICARRIER RECEIVERS AND METHODS FOR SEPARATING TRANSMITTED SIGNALS IN A MULTIPLE ANTENNA SYSTEM



(57) Abstract: A multicarrier receiver includes spatial channel decouplers to separate signal components that are separately transmitted with more than one transmit antenna or over more than one spatial channel. The decouplers separate the transmitted signals based on a QR factorization of the channel estimate matrix.

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# MULTICARRIER RECEIVERS AND METHODS FOR SEPARATING TRANSMITTED SIGNALS IN A MULTIPLE ANTENNA SYSTEM

### Cross-Reference to Related Applications

5           This application claims the benefit of priority to U.S. Provisional Patent Application Serial No. 60/536,071, filed January 12, 2004, and U.S. Patent Application Serial No. 10/814,095 filed March 30, 2004, which are incorporated herein by reference.

10 Technical Field

Embodiments of the present invention pertain to electronic communications and in some embodiments, to multicarrier communications.

## Background

15        Wireless signals that are transmitted with more than one transmit  
antenna or over more than one non-orthogonal spatial channel on the same  
frequencies are combined in the communication channel. These signals should  
be separated again in the receiver for proper decoding. Some conventional  
receivers employ maximum-likelihood (ML) decoding to separate different  
20        transmitted signals. These maximum-likelihood decoding techniques are  
computationally expensive, even for signals transmitted with as few as two  
transmit antennas. Some other conventional receivers employ minimum-mean-  
squared error (MMSE) decoding to reduce the amount of computations for  
separating different transmitted signals. In many situations, MMSE decoding  
25        techniques fall significantly short of ML decoding techniques and may result in a  
significant reduction in performance, especially for a practical wireless  
communication system.

Thus, there are general needs for receivers and methods of decoding signals transmitted with more than transmit antenna or over more than one spatial channel. There are also needs for receivers and methods of decoding

signals with reduced processing. There are also needs for receivers and methods of decoding signals with better performance.

#### Brief Description of the Drawings

5           The appended claims are directed to some of the various embodiments of the present invention. However, the detailed description presents a more complete understanding of embodiments of the present invention when considered in connection with the figures, wherein like reference numbers refer to similar items throughout the figures and:

10           FIG. 1 is a block diagram of a receiver in accordance with some embodiments of the present invention;

            FIG. 2 illustrates a block diagram of a spatial channel decoupler in accordance with some embodiments of the present invention; and

            FIG. 3 illustrates a signal component separation procedure in accordance  
15   with some embodiments of the present invention.

#### Detailed Description

            The following description and the drawings illustrate specific embodiments of the invention sufficiently to enable those skilled in the art to  
20   practice them. Other embodiments may incorporate structural, logical, electrical, process, and other changes. Examples merely typify possible variations. Individual components and functions are optional unless explicitly required, and the sequence of operations may vary. Portions and features of some  
embodiments may be included in or substituted for those of others. The scope of  
25   embodiments of the invention encompasses the full ambit of the claims and all available equivalents of those claims. Such embodiments of the invention may be referred to, individually or collectively, herein by the term "invention" merely for convenience and without intending to voluntarily limit the scope of this

application to any single invention or inventive concept if more than one is in fact disclosed.

FIG. 1 is a block diagram of a receiver in accordance with some embodiments of the present invention. Receiver 100 may be part of a wireless communication device and may receive multicarrier communication signals over a multicarrier communication channel having more than one spatial channel within a predetermined frequency spectrum. In some embodiments, the multicarrier communication signals may be orthogonal frequency division multiplexed (OFDM) communication signals or discrete multi-tone modulated (DMT) signals, although the scope of the invention is not limited in this respect. The multicarrier communication channel may comprise a plurality of orthogonal subcarriers. In some embodiments, the orthogonal subcarriers of the multicarrier channel may be closely spaced subcarriers, such as OFDM subcarriers. To achieve orthogonality between the closely spaced subcarriers, the subcarriers of a particular channel may have been transmitted with a null at substantially a center frequency of the other subcarriers of that channel, although the scope of the invention is not limited in this respect.

In some embodiments, a transmitting station may utilize more than one spatially-diverse transmit antenna to “divide” the channel into one or more spatial channels. In some embodiments, each transmit antenna may define one spatial transmit channel. In other embodiments, the transmitting station may employ beamforming techniques to “divide” the channel into spatial channels. In these embodiments, each spatial channel may be used to communicate separate or independent data streams on the same subcarriers as the other spatial channels, allowing the communication of additional data without an increase in frequency bandwidth. The use of spatial channels may take advantage of the multipath characteristics of the channel. In some embodiments, the spatial channels may be non-orthogonal channels, although the scope of the invention is not limited in this respect.

In some embodiments, a multicarrier symbol, such as an OFDM symbol, may be viewed as the combination of the symbols modulated on the individual subcarriers of the spatial channels. Because of the range in the number of bits that may be modulated per symbol on each subcarrier and the variable number of spatial channels that may be used, the number of bits per symbol may vary greatly. In some embodiments, subcarrier modulation assignments may be based on channel conditions, such as a signal to interference and noise ratio (SINR) for the particular subcarrier in a particular spatial channel. In some embodiments, the subcarrier modulation assignments may be selected by the receiving station and provided to the transmitting station, although the scope of the invention is not limited in this respect. In some embodiments, higher subcarrier modulation assignments (e.g., more bits per symbol) may be used for subcarriers having better SINRs.

Receiver 100 may comprise one or more of spatially diverse antennas 102 and radio-frequency (RF) circuitry 104 to receive a multicarrier symbol over a plurality of spatial channels. Receiver 100 may also comprise fast Fourier transform (FFT) circuitry 106 to generate frequency domain representations 107 of the multicarrier symbol received over the subcarriers. Receiver 100 may also comprise subcarrier signal collectors 108 to collect signals of an associated subcarrier from each of FFT circuitry 106 to generate complex signals 109. Each of complex signals 109 may include signal components that were separately transmitted on the same subcarrier frequency with more than one transmit antenna or over more than one spatial channel. For proper decoding, the signal components of complex signals 109 should be separated.

Receiver 100 may also comprise spatial channel decouplers 110 to separate from complex signals 109, the signal components separately transmitted with more than one transmit antenna or over more than one spatial channel. Spatial channel decouplers 110 may use channel estimates 119 and may generate transmitted complex signals 111 having components (e.g.,  $x_1$  and  $x_2$ ) for each

transmit antenna or each transmit spatial channel. In some embodiments, receiver 100 may be provided the number of transmit antennas used by a transmitting station as part of channelization information of a packet.

Receiver 100 may also comprise demapper 112 to demap symbols from  
5 each subcarrier to bits based on the modulation order used in transmission. In some embodiments, demapper 112 may generate bits from signals 111 for each transmit antenna or each transmit spatial channel. In some embodiments, demapper 112 may include functionally separate demapper circuitry for each subcarrier of the channel. In some embodiments, demapper 112 may include  
10 functionally separate demapper circuitry for each transmit antenna or each transmit spatial channel, although the scope of the invention is not limited in this respect.

Receiver 100 may also comprise deinterleaver 114 to perform a deinterleaving operation on blocks of bits 113 provided by demapper 112.  
15 Receiver 100 may also comprise decoder 116 to decode demapped blocks of bits 115 to generate decoded bit sequence 117. Decoder 116 may be an error correcting decoder or convolutional decoder, although the scope of the invention is not limited in this respect.

In some embodiments, receiver 100 may include an RF chain for each  
20 spatial channel. The RF chain may comprise one of RF circuitry 104 and an associated one of FFT circuitry 106 for each spatial channel. FIG. 1 illustrates a receiver with four RF chains; however the scope of the invention is not limited in this respect. In some embodiments, antennas 102 may be spatially diverse antennas and each may be associated with a spatial channel. Although one of  
25 antennas 102 is illustrated for each RF chain, this is not a requirement.

In some other embodiments, beamforming techniques may be employed to provide more than one spatial channel in a multicarrier communication channel. For example, instead of one antenna 102 for each RF chain (as

illustrated), a beamformer (not illustrated) may be provided between one or more antennas 102 and RF circuitry 104.

Subcarrier signal collectors 108 may be associated with particular subcarrier frequencies rather than spatial channels so that any one signal collector 108 may collect signals for corresponding subcarriers (i.e., the same subcarrier frequency) of each of the spatial channels. Multiple symbols for each subcarrier may be collected and/or combined by each signal collector 108 to generate complex received signals 109 having signal components from each transmitted signal.

10 In accordance with some embodiments of the present invention, spatial channel decouplers 110 may separate signal components (e.g.,  $x_1$  and  $x_2$ ) of a multicarrier signal transmitted over a plurality of spatial channels. In these embodiments, spatial channel decouplers 110 may perform a decomposition on a channel estimate matrix (H) to generate a unitary orthogonal matrix (Q) and an upper triangular matrix (R). The channel estimate matrix (H) may comprise a matrix of channel estimates 119 (i.e., the channel response) at a particular subcarrier frequency for the different spatial channels between receiver 100 and a transmitting station. In some embodiments that comprise M transmit antennas and N receiver antennas, the channel estimate matrix may be an  $M \times N$  matrix. In 15 some embodiments, M and N may range from two to ten, although the scope of the invention is not limited in this respect.

In some embodiments that receive signals transmitted over two transmit spatial channels, spatial channel decouplers 110 may multiply received complex signals (y) 109 by a conjugate transpose ( $Q^*$ ) of the unitary orthogonal matrix (Q) to generate a z-vector. In some embodiments, spatial channel decouplers 110 25 may estimate transmitted complex signal components (e.g.,  $x_1$  and  $x_2$ ) of the multicarrier signal from the upper triangular matrix (R), and components of the z vector. In some embodiments, each of spatial channel decouplers 110 may estimate a second transmitted complex signal ( $x_2$ ) representing a second signal



component on a particular subcarrier frequency of the transmitted multicarrier signal solely from a portion of the upper triangular matrix (R) and a second component of the z-vector (z2). In some embodiments, each of spatial channel decouplers 110 may then estimate a first transmitted complex signal (x1) representing a first signal component of the transmitted multicarrier signal based on the estimated transmitted complex signal (x2), a portion of the upper triangular matrix (R), and a first component of the z vector (z1). This is further described below.

In these embodiments, received complex signals (y) 109 may be represented by the following equation:

Eq. 1:  $y = Hx + v$ , where H represents the channel estimate matrix, x represents the transmitted complex signal and v represents noise.

The channel estimate matrix (H) may be represented by the following equation:

Eq. 2:  $H = QR$ , where Q represents the unitary orthogonal matrix and R represents the upper triangular matrix of the QR decomposition of matrix H.

By substituting QR for H in Eq. 1 and multiplying both sides of Eq. 1 by a conjugate transpose ( $Q^*$ ) of the unitary orthogonal matrix (Q), the following equation may result:

Eq. 3:  $z = Rx + n$ , where z is the received complex signals (y) multiplied by conjugate transpose ( $Q^*$ ), R is the upper triangular matrix, x represents the transmitted complex signals, and n (a vector) represents noise. The conjugate transpose of the unitary orthogonal matrix (Q) may be a hermitian of the unitary orthogonal matrix and may be viewed as a hermitian matrix ( $Q^*$ ).

In some embodiments in which either two transmit antennas or two spatial transmit channels are used for transmitting signals, as shown below, the vector z may have first and second components, the matrix R may be viewed as an upper diagonal matrix, and the transmitted complex signal (x) may have first and second components, as illustrated below.

$$z = \begin{bmatrix} z1 \\ z2 \end{bmatrix}, R = \begin{bmatrix} r1 & r3 \\ 0 & r2 \end{bmatrix}, x = \begin{bmatrix} x1 \\ x2 \end{bmatrix}$$

In these embodiments, the following equations may be generated from  
 5 the above:

$$\text{Eq. 4: } z1 = r1x1 + r3x2 + n1$$

$$\text{Eq. 5: } z2 = 0x1 + r2x2 + n2$$

The second transmitted complex signal (x2) (i.e., the signal transmitted  
 by a second transmit antenna and/or over a second spatial channel) may be  
 10 estimated using Eq. 5. Based on Eq. 5, the second transmitted complex signal  
 (x2) may be an interference-free estimate (i.e., without contribution from x1). In  
 some embodiments, the second transmitted complex signal (x2) may be  
 estimated by scalar methods such as equalization and slicing, although the scope  
 of the invention is not limited in this respect. The second transmitted complex  
 15 signal (x2) may be viewed as lying on a grid of real and imaginary signals.

Once the second transmitted complex signal (x2) is estimated from Eq. 5,  
 Eq. 4 may be used to estimate the first transmitted complex signal (x1) (i.e., the  
 signal transmitted by the first transmit antenna and/or over a first spatial  
 channel). In some embodiments, the first transmitted complex signal (x1) may be  
 20 estimated by nulling and canceling techniques.

In some embodiments, each of spatial channel decouplers 110 may  
 provide the estimated first and second transmitted complex signals (x1 and x2)  
 111 for a particular subcarrier frequency as its output. In some embodiments, the  
 first and second transmitted complex signals (x1 and x2) may be complex  
 25 quadrature-amplitude-modulated (QAM) signals, although the scope of the  
 present invention is not limited in this respect.

Estimates for the first and second transmitted complex signals (x1 and  
 x2) may have been affected by noise. In some embodiments, spatial channel

decouplers 110 may perform further operations to revise the estimates for the first and second transmitted complex signals (x1 and x2). In these embodiments, spatial channel decouplers 110 may estimate nearest neighbors of the transmitted complex signal (x2) and may also re-estimate the transmitted complex signal (x1) for each of the nearest neighbors of the transmitted complex signal (x2). Spatial channel decouplers 110 may also compute Euclidian distances between components of the z vector and H\* (i.e., the initially computed transmitted complex signal (x1) and the estimated transmitted complex signal (x2)), and between components of the received z-vector and H\* (i.e., the re-estimated transmitted complex signals (x1) and the corresponding neighbors of transmitted complex signal (x2)). In these embodiments, spatial channel decouplers 110 may select the pair of estimates for the transmitted complex signals (including the neighbors and estimates based on the neighbors) with the lowest computed distance. The selected pair may be a revised estimate for the first and second transmitted complex signals (x1 and x2). In these embodiments, spatial channel decouplers 110 may provide the revised estimate for the first and second transmitted complex signals (x1 and x2) 111 as its output. In some embodiments, crosstalk (i.e., interference) between non-orthogonal spatial channels may be substantially decoupled or separated from received complex signals (y) 109.

In some embodiments, the nearest neighbors (xn) of the estimated second transmitted complex signal (x2) may be calculated by determining other closest points to the second component of the z vector (z2) on a grid of real and imaginary signals representing the estimated second transmitted complex signal (x2). In this way, the computed Euclidian distance using the neighbors may be only slightly larger than the Euclidian distance using the original estimate (e.g.,  $\|z_2 - r_2 * \text{neighbor of } x_2\|$  may be slightly larger than  $\|z_2 - r_2 * x_2\|$ ).

In some embodiments, the Euclidian distance may be computed by the following equation:

$$\text{Eq. 6: } E_0 = \|y - Hx\|^2$$

Although Eqs. 1 through 6 are depicted for some embodiments of the present invention in which signals are transmitted with either two transmit antennas or employing two spatial channels, these equations may be extended for embodiments of the present invention in which signals are transmitted with more  
5 than two antennas or over more than two spatial channels.

In some embodiments, the channel estimate matrix (H) may comprise a matrix of channel estimates at a specific subcarrier frequency for the spatial channels between receiver 100 and a transmitting station. In some embodiments, the plurality of spatial channels may comprise four spatial channels defined by  
10 paths between each transmit antenna of a pair of transmit antennas and each receive antenna of a pair receive antennas. In this case, the channel estimate matrix (H) may comprise a 2x2 matrix. In some other embodiments, the plurality of spatial channels may comprise up to sixteen spatial channels defined by paths between each transmit antenna of up to four transmit antennas and each receive  
15 antenna of up to four receive antennas. In this case, the channel estimate matrix (H) may comprise up to a 4x4 matrix. Embodiments of the present invention are also suitable for almost any number spatial channels.

In some embodiments, the transmitted multicarrier signals may comprise two multicarrier signal components of a single multicarrier symbol transmitted  
20 substantially simultaneously over the plurality of spatial channels. Each multicarrier signal component may comprise a plurality of symbol-modulated subcarriers transmitted on orthogonal symbol-modulated subcarriers of the multicarrier channel. The spatial channels may be non-orthogonal channels, each employing symbol-modulated subcarriers of the same frequencies (i.e., tones of  
25 the same frequency).

In some embodiments, each of spatial channel decouplers 110 may provide the estimated first and second transmitted complex signals (x1 and x2) 111 as its output for one subcarrier of the plurality of subcarriers comprising the multicarrier channel. In some embodiments, channel estimates 119 may be

channel estimates for each one of the subcarrier frequencies, and accordingly, there may be a channel matrix (H) for each subcarrier frequency.

In some embodiments, demapper 112 may perform quadrature amplitude modulation (QAM) demapping on the first transmitted complex signal (x1) and the second transmitted complex signal (x2) to generate bits 113 corresponding to the first and second transmitted complex signals (x1 and x2) for each subcarrier. Deinterleaver 114 may deinterleave bits 113 to generate deinterleaved bits 115. Decoder 116 may decode (and in some embodiments combine) deinterleaved bits 115 corresponding to the first and second transmitted complex signals (x1 and x2) to generate combined bit stream 117 from all subcarriers representing a transmitted multicarrier symbol. In other embodiments, the combining of bits separated from the separately transmitted signals may be performed by deinterleaver 114, although the scope of the invention is not limited in this respect.

In some embodiments, the frequency spectrums for the multicarrier channel may comprise either a 5 GHz frequency spectrum or a 2.4 GHz frequency spectrum. In these embodiments, the 5 GHz frequency spectrum may include frequencies ranging from approximately 4.9 to 5.9 GHz, and the 2.4 GHz spectrum may include frequencies ranging from approximately 2.3 to 2.5 GHz, although the scope of the invention is not limited in this respect, as other frequency spectrums are also equally suitable.

In some embodiments, receiver 100 may be a personal digital assistant (PDA), a laptop or portable computer with wireless communication capability, a web tablet, a wireless telephone, a wireless headset, a pager, an instant messaging device, a digital camera, an access point or other device that may receive and/or transmit information wirelessly. In some embodiments, RF circuitry 104 may receive RF communications in accordance with specific communication standards, such as the Institute of Electrical and Electronics Engineers (IEEE) standards including IEEE 802.11(a), 802.11(b), 802.11(g/h),

802.11(n) and/or 802.16 standards for wireless local area networks, although receiver 100 may also be suitable to receive communications in accordance with other techniques including the Digital Video Broadcasting Terrestrial (DVB-T) broadcasting standard, and the High performance radio Local Area Network (HiperLAN) standard.

Although some embodiments of the present invention are discussed in the exemplary context of an 802.11x implementation (e.g., 802.11a, 802.11g, 802.11 HT, etc.), the claims are not so limited. Indeed, embodiments of the present invention may well be implemented as part of any wireless system using multicarrier wireless communication channels (e.g., orthogonal frequency-division multiplexing (OFDM), discrete multi-tone modulation (DMT), etc.), such as may be used within, without limitation, a wireless personal area network (WPAN), a wireless local area network (WLAN), a wireless metropolitan area network (WMAN), a wireless wide area network (WWAN), a cellular network, a third generation (3G) network, a fourth generation (4G) network, a universal mobile telephone system (UMTS), and the like communication systems.

In some embodiments, antennas 102 may comprise one or more of a directional or omnidirectional antenna, including, for example, a dipole antenna, a monopole antenna, a loop antenna, a microstrip antenna or other type of antenna suitable for reception and/or transmission of RF signals which may be processed by RF circuitry 104.

In accordance with some embodiments, the subcarriers of each spatial channel may have been individually symbol-modulated in accordance with individual subcarrier modulation assignments. This may be referred to as adaptive bit loading (ABL). Accordingly, a variable number of bits may be represented by a symbol-modulated on a subcarrier. The modulation assignments for each spatial channel may be based on the channel characteristics or channel conditions for that spatial channel, although the scope of the invention is not limited in this respect. In some embodiments, the subcarrier modulation

assignments may range from zero bits per symbol to up to ten or more bits per symbol. In terms of modulation levels, the subcarrier modulation assignments may comprise binary phase shift keying (BPSK), which communicates one bit per symbol, quadrature phase shift keying (QPSK), which communicates two  
5 bits per symbol, 8PSK, which communicates three bits per symbol, 16-quadrature amplitude modulation (16-QAM), which communicates four bits per symbol, 32-QAM, which communicates five bits per symbol, 64-QAM, which communicates six bits per symbol, 128-QAM, which communicates seven bits per symbol, and 256-QAM, which communicates eight bits per symbol.

10 Subcarrier modulation assignments with higher data communication rates per subcarrier (e.g., ten bits) may also be used. In other embodiments, the same subcarrier modulation assignment may be employed on most or all subcarriers of a multicarrier signal.

Although receiver 100 is illustrated as having several separate functional  
15 elements, one or more of the functional elements may be combined and may be implemented by combinations of software-configured elements, such as processing elements including digital signal processors (DSPs), and/or other hardware elements. For example, the elements illustrated may comprise one or more microprocessors, DSPs, application specific integrated circuits (ASICs),  
20 and combinations of various hardware and logic circuitry for performing at least the functions described herein.

FIG. 2 illustrates a block diagram of a spatial channel decoupler in accordance with some embodiments of the present invention. Spatial channel decoupler 200 may be suitable for use as one of spatial channel decouplers 110  
25 (FIG. 1), although other spatial channel decoupler configurations may also be suitable. From received signals ( $y_1$  and  $y_2$ ) 201, spatial channel decoupler 200 may separate signal components that were separately transmitted with more than one transmit antenna or over more than one spatial channel.

Spatial channel decoupler 200 may comprise decomposer 202 to perform a QR decomposition on a channel estimate matrix (H). The channel matrix may be generated from channel estimates 219 for the appropriate subcarrier frequency. Decomposer 202 may generate a unitary orthogonal matrix (Q) and  
5 upper triangular matrix (R) 205 from the channel estimate matrix (H).

Spatial channel decoupler 200 may also comprise multiplier 204 to multiply received complex signals ( $y_1$  and  $y_2$ ) 201 by conjugate transpose ( $Q^*$ ) 203 of the unitary orthogonal matrix (Q) to generate z-vector 207. In some embodiments, decomposer 202 may generate conjugate transpose ( $Q^*$ ) 203,  
10 while in other embodiments, multiplier 204 may generate conjugate transpose ( $Q^*$ ) from the unitary orthogonal matrix (Q).

Spatial channel decoupler 200 may also comprise transmitted signal initial estimator 208 to estimate transmitted complex signal components ( $x_1$  and  $x_2$ ) of the multicarrier signal from the upper triangular matrix (R), and  
15 components of the z vector. In some embodiments, signal initial estimator 208 may generate an initial estimate of a transmitted complex signal, such as second transmitted complex signal ( $x_2$ ) representing a second signal component of the multicarrier signal. Estimator 208 may use the upper triangular matrix (R) 205 and components of z-vector ( $z_2$ ) 207. Transmitted signal initial estimator 208  
20 may also estimate other transmitted complex signals, such as first transmitted complex signal ( $x_1$ ) representing a first signal component of the multicarrier signal. Estimator 208 may use the previously estimated transmitted complex signal, such as second transmitted complex signal ( $x_2$ ), upper triangular matrix (R) 205, and components of z-vector ( $z_1$ ) 207.

25 In some embodiments, spatial channel decoupler 200 may also comprise nearest neighbor calculator 210 to estimate nearest neighbors. In some embodiments, nearest neighbor calculator 210 may estimate nearest neighbors of the transmitted complex signal ( $x_2$ ) and may also re-estimate the transmitted complex signal ( $x_1$ ) for each of the nearest neighbors estimated for the



transmitted complex signal ( $x_2$ ). In some embodiments, the nearest neighbors may be based on Euclidian distances computed between components of the  $z$  vector and  $H^*$  (i.e., the initially computed transmitted complex signal ( $x_1$ ) and the estimated transmitted complex signal ( $x_2$ )), and between components of the  
5 received  $z$ -vector and  $H^*$  (i.e., the re-estimated transmitted complex signals ( $x_1$ ) and the corresponding neighbors of transmitted complex signal ( $x_2$ )).

In some embodiments, Euclidian distance calculator 212 may be used to compute Euclidian distances. In the case of two transmit antennas or two spatial channels, Euclidian distance calculator 212 may compute Euclidian distances  
10 between the estimated first transmitted complex signals ( $x_1$ ) and the initially estimated second transmitted complex signal ( $x_2$ ), and between each re-estimated first transmitted complex signals ( $x_1$ ) (e.g., corresponding to the nearest neighbor of second transmitted complex signals ( $x_2$ )) and each neighbor of the initially estimated second transmitted complex signals ( $x_2$ ).

15 In some embodiments, spatial channel decoupler 200 may also comprise selector 214 to select a set of transmitted complex signals associated with a lowest computed Euclidian distance. In the case of two transmit antennas or two spatial channels, a selected pair may corresponding to the first and second transmitted complex signal components ( $x_1$  and  $x_2$ ) 215 and may be provided as  
20 an output.

Although spatial channel decoupler 200 is illustrated for embodiments of the present invention that receive signals transmitted over two spatial channels or with two transmit antennas, the scope of the present invention is not limited in this respect. In some embodiments, spatial channel decoder 200 may be suitable  
25 to decouple signals transmitted using more than two spatial channels or more than two transmit antennas.

Although spatial channel decoupler 200 is illustrated as having several separate functional elements, one or more of the functional elements may be combined and may be implemented by combinations of software-configured

elements, such as processing elements including digital signal processors (DSPs), and/or other hardware elements. For example, the elements illustrated may comprise one or more microprocessors, DSPs, application specific integrated circuits (ASICs), and combinations of various hardware and logic  
5 circuitry for performing at least the functions described herein.

FIG. 3 illustrates a signal component separation procedure in accordance with some embodiments of the present invention. Procedure 300 may be performed by a receiver, such as receiver 100 (FIG. 1), although other receivers and receiving stations may also perform procedure 300. In some embodiments,  
10 procedure 300 may be performed by one or more spatial channel decouples, such as spatial channel decoupler 200 (FIG. 2). Procedure 300 may be used to separate signal components of a multicarrier signal transmitted using a plurality of transmit antennas or over a plurality of spatial channels.

Operation 302 comprises performing a decomposition on a channel  
15 estimate matrix (H) to generate a unitary orthogonal matrix (Q) and an upper triangular matrix (R). In some embodiments, operation 302 may be performed by decomposer 202 (FIG. 2), although the scope of the invention is not limited in this respect.

Operation 304 comprises multiplying received complex signals (y) by a  
20 conjugate transpose ( $Q^*$ ) of the unitary orthogonal matrix (Q) to generate a z-vector. In some embodiments, operation 304 may be performed by multiplier 204 (FIG. 2), although the scope of the invention is not limited in this respect.

Operation 306 comprises estimating transmitted complex signal components (e.g.,  $x_1$  and  $x_2$ ) of the multicarrier signal from the upper triangular  
25 matrix (R), the upper triangular matrix (R), and components of the z vector. In some embodiments, operation 306 comprises estimated a transmitted complex signal, such as second transmitted complex signal ( $x_2$ ) representing a second signal component of the multicarrier signal. Operation 306 may utilize a portion of the upper triangular matrix (R) and one or more components of the z-vector.

In some embodiments, operation 306 may be performed by signal estimator 208 (FIG. 2), although the scope of the invention is not limited in this respect.

Operation 308 comprises estimating nearest neighbors of the transmitted complex signal estimated in operation 306. In some embodiments, operation 308  
5 may be performed by neighbor calculator 210 (FIG. 2), although the scope of the invention is not limited in this respect.

Operation 310 comprises estimating other transmitted complex signals, such as a first transmitted complex signal ( $x_1$ ) representing a first signal component of the multicarrier signal. In some embodiments, operation 310 may  
10 utilize the previously or initially estimated neighbors of the previously or initially estimated transmitted complex symbol, such as second transmitted complex signal ( $x_2$ ), the upper triangular matrix ( $R$ ), and components of the  $z$  vector. In some embodiments, operation 310 may be performed by signal estimator 208 (FIG. 2), although the scope of the invention is not limited in this respect.

15 In some embodiments, the estimated transmitted complex signals may be provided for subsequent QAM demapping. In some other embodiments, procedure 300 may also include performing operations 312 through 314 to further revise the estimates.

Operation 312 comprises computing Euclidian distances between the  
20 transmitted complex signal estimated in operation 306 and the transmitted complex signals estimated in operation 310. Operation 312 may also comprise computing Euclidian distances between the transmitted complex signal estimated in operation 306 and each neighbor estimated in operation 308. In some embodiments, operation 312 may be performed by distance calculator 212 (FIG.  
25 2), although the scope of the invention is not limited in this respect.

Operation 313 comprises repeating operations 308 through 312 for other neighbors as part of a recursive process. For example, each repetition of operations 308 through 312 may use the one or more of the neighbors from the prior operations. In this way, in the case of two transmit and two receive

antennas, an estimate for first transmitted complex signal ( $x_1$ ) may be generated for each neighbor of second transmitted complex signal ( $x_2$ ). In some embodiments, operation 313 may be performed by signal estimator 208 (FIG. 2), neighbor calculator 210 (FIG. 2) and distance calculator 212 (FIG. 2) in  
5 conjunction with a system controller (not separately illustrated in FIG. 2), although the scope of the invention is not limited in this respect.

Operation 314 comprises selecting a set of transmitted complex signals associated with a lowest computed distance. In the case of two transmit antennas or two transmit spatial channels, the selected set may correspond to the first and  
10 second transmitted complex signal components ( $x_1$  and  $x_2$ ). In some embodiments, operation 314 may be performed by selector 214 (FIG. 2), although the scope of the invention is not limited in this respect.

In multicarrier embodiments, operation 316 comprises repeating operations 302 through 314 for each subcarrier frequency. In some embodiments,  
15 operations 302 through 314 may be performed concurrently for each subcarrier frequency, such as by a plurality of spatial channel decoders, such as spatial channel decoders 110 (FIG. 1). In some embodiments, operation 316 may be performed by a separate one of spatial channel decouplers, such as spatial channel decouplers 110 (FIG. 1) for each subcarrier of the multicarrier  
20 communication channel, although the scope of the invention is not limited in this respect. In these embodiments, each spatial channel decoupler may perform operations 302 through 314 in parallel for a particular subcarrier frequency.

Although the individual operations of procedure 300 are illustrated and described as separate operations, one or more of the individual operations may  
25 be performed concurrently, and nothing requires that the operations be performed in the order illustrated.

Unless specifically stated otherwise, terms such as processing, computing, calculating, determining, displaying, or the like, may refer to an action and/or process of one or more processing or computing systems or similar

devices that may manipulate and transform data represented as physical (e.g., electronic) quantities within a processing system's registers and memory into other data similarly represented as physical quantities within the processing system's registers or memories, or other such information storage, transmission  
5 or display devices. Furthermore, as used herein, computing device includes one or more processing elements coupled with computer-readable memory that may be volatile or non-volatile memory or a combination thereof. Moreover, as used herein, data refers to one or more storage data elements, which can include portions of files, a single file, a file extent, a database, a storage device partition,  
10 a volume, sets of volumes and the like. The data need not reside on a single storage device and may span multiple storage devices.

Some embodiments of the invention may be implemented in one or a combination of hardware, firmware and software. Embodiments of the invention may also be implemented as instructions stored on a machine-readable medium,  
15 which may be read and executed by at least one processor to perform the operations described herein. A machine-readable medium may include any mechanism for storing or transmitting information in a form readable by a machine (e.g., a computer). For example, a machine-readable medium may include read-only memory (ROM), random-access memory (RAM), magnetic  
20 disk storage media, optical storage media, flash-memory devices, electrical, optical, acoustical or other form of propagated signals (e.g., carrier waves, infrared signals, digital signals, etc.), and others.

The Abstract is provided to comply with 37 C.F.R. Section 1.72(b) requiring an abstract that will allow the reader to ascertain the nature and gist of  
25 the technical disclosure. It is submitted with the understanding that it will not be used to limit or interpret the scope or meaning of the claims.

In the foregoing detailed description, various features are occasionally grouped together in a single embodiment for the purpose of streamlining the disclosure. This method of disclosure is not to be interpreted as reflecting an

intention that the claimed embodiments of the subject matter require more features than are expressly recited in each claim. Rather, as the following claims reflect, invention lies in less than all features of a single disclosed embodiment. Thus the following claims are hereby incorporated into the detailed description,  
5 with each claim standing on its own as a separate preferred embodiment.

## CLAIMS

What is claimed is:

1. A method of separating signal components comprising:
  - 5 performing a decomposition on a channel estimate matrix to generate a unitary orthogonal matrix and an upper triangular matrix;
  - multiplying received complex signals by a conjugate transpose of the unitary orthogonal matrix to generate a z-vector; and
  - estimating transmitted complex signal components of a multicarrier
  - 10 signal from the upper triangular matrix, the upper triangular matrix, and components of the z vector.
2. The method of claim 1 wherein estimating comprises:
  - estimating a second transmitted complex signal representing a second
  - signal component of the multicarrier signal from the upper triangular matrix and
  - 15 a second component of the z-vector; and
  - estimating a first transmitted complex signal representing a first signal component of the multicarrier signal based on the estimated second transmitted complex signal, the upper triangular matrix, and a first component of the z
  - vector.
- 20 3. The method of claim 2 further comprising:
  - estimating nearest neighbors of the estimated second transmitted complex
  - signal;
  - re-estimating the first transmitting complex signal for each of the nearest
  - neighbors of the second transmitted complex signal;
  - 25 computing Euclidian distances between the first transmitted complex
  - signal and the second transmitted complex signal, and between the re-estimated

first transmitted complex signals and each neighbor of the second transmitted complex signal; and

selecting a pair of signal estimates associated with the lowest computed distance, the selected pair corresponding to final estimates of the first and second  
5 transmitted complex signals.

4. The method of claim 3 wherein the transmitted complex signal components were transmitted separately over more than one non-orthogonal spatial channel using more than one corresponding transmit antennas, and  
wherein the final estimates of the first and second transmitted complex  
10 signals are substantially decoupled from crosstalk between the non-orthogonal spatial channels.

5. The method of claim 1 wherein the z-vector is equal to a noise factor plus the upper triangular matrix multiplied by an x-vector,  
wherein components of the x-vector represent individual complex signal  
15 components transmitted over corresponding individual spatial channels, and  
wherein the conjugate transpose of the unitary orthogonal matrix is a hermitian of the unitary orthogonal matrix comprising a hermitian matrix.

6. The method of claim 2 wherein estimating the second transmitted complex signal comprises generating a substantially interference-free estimate of  
20 the second transmitted complex signal, the interference-free estimate being substantially free from contributions of the first transmitted complex signal.

7. The method of claim 4 wherein the channel estimate matrix comprises a matrix of channel estimates for the plurality of spatial channels between a receiving station and a transmitting station.



8. The method of claim 7 wherein the plurality of spatial channels comprise spatial channels defined by communication paths between a pair of transmit antennas and a pair receive antennas, and wherein the channel estimate matrix comprises a 2x2 matrix.

5           9. The method of claim 7 wherein the plurality of spatial channels comprise spatial channels defined by:  
          either two transmit antennas or a single transmit antenna coupled to the transmitting station employing beamforming techniques to define two transmit spatial channels;  
10          either two receive antennas or a single receive antenna coupled to the receiving station employing beamforming techniques to define two receive spatial channels, and  
          wherein the channel estimate matrix comprises a 2x2 matrix of channel estimates for the spatial channels.

15           10. The method of claim 2 wherein the multicarrier signal comprises two orthogonal frequency division multiplexed signal components of a single orthogonal frequency division multiplexed symbol transmitted substantially simultaneously over the plurality of spatial channels,  
          wherein each orthogonal frequency division multiplexed signal  
20          component comprises a plurality of symbol-modulated subcarriers transmitted on orthogonal symbol-modulated subcarriers, and  
          wherein the spatial channels are non-orthogonal channels, each employing symbol-modulated subcarriers of the same frequencies.

          11. The method of claim 10 wherein the performing, the multiplying, the  
25          estimating a second transmitted complex signal, and the estimating a first transmitted complex symbol are performed for a first subcarrier of the plurality of symbol-modulated subcarriers, and

wherein the method further comprises repeating the performing, the multiplying, the estimating a second transmitted complex signal, and the estimating a first transmitted complex symbol for other subcarriers of the plurality, and

- 5            wherein performing the decomposition comprises performing a decomposition of a channel matrix of channel estimates for each of the subcarriers of the plurality of subcarriers.

12. The method of claim 10 wherein each of the two orthogonal frequency division multiplexed signal components is to have been modulated  
10        with separate data symbols of a single transmitted orthogonal frequency division multiplexed symbol.

13. The method of claim 12 further comprising:

- for each subcarrier, performing a quadrature amplitude modulation demapping on the first transmitted complex signal and the second transmitted  
15        complex signal to generate corresponding first and second bits for each subcarrier; and

              deinterleaving and decoding the first and second bits from the subcarriers to generate a combined bit stream representing the single transmitted orthogonal frequency division multiplexed symbol.

- 20            14. A spatial channel decoupler comprising:

              a decomposer to perform a decomposition on a channel estimate matrix to generate a unitary orthogonal matrix and an upper triangular matrix;

              a multiplier to multiply received complex signals by a conjugate transpose of the unitary orthogonal matrix to generate a z-vector; and

- 25            a transmitted signal estimator to estimate transmitted complex signal components of a multicarrier signal from the upper triangular matrix, the upper triangular matrix, and components of the z vector.

15. The decoupler of claim 14 wherein the transmitted signal estimator estimates a second transmitted complex signal representing a second signal component of the multicarrier signal from the upper triangular matrix and a second component of the z-vector, and

5        wherein the transmitted signal estimator further estimates a first transmitted complex signal representing a first signal component of the multicarrier signal based on the estimated second transmitted complex signal, the upper triangular matrix, and a first component of the z vector.

16. The decoupler of claim 15 further comprising:

10        a neighbor calculator to estimate nearest neighbors of the second transmitted complex signal;

         a distance calculator to compute Euclidian distances; and  
         a selector,

         wherein the transmitted signal estimator re-estimates the first transmitted  
15        complex signal for each of the nearest neighbors of the second transmitted complex signal,

         wherein the distance calculator computes Euclidian distances between the first transmitted complex signal and the second transmitted complex signal, and between the re-estimated first transmitted complex signals and each  
20        neighbor of the second transmitted complex signal, and

         wherein the selector selects a pair of signal estimates associated with the lowest computed distance, the selected pair corresponding to a final estimate of the first and second transmitted complex signals.

17. The decoupler of claim 16 wherein the transmitted complex signal  
25        components were transmitted separately over more than one non-orthogonal spatial channel using more than one corresponding transmit antennas, and

         wherein the final estimate of the first and second transmitted complex signals is substantially decoupled from crosstalk between non-orthogonal spatial channels.

18. The decoupler of claim 14 wherein the z-vector is equal to a noise factor plus the upper triangular matrix multiplied by an x-vector,  
wherein components of the x-vector represent individual complex signal components transmitted over corresponding individual spatial channels, and  
5 wherein the conjugate transpose of the unitary orthogonal matrix is a hermitian of the unitary orthogonal matrix comprising a hermitian matrix.

19. The decoupler of claim 15 wherein the transmitted signal estimator is to generate an interference-free estimate of the second transmitted complex signal, the interference-free estimate being substantially free from contributions  
10 of the first transmitted complex signal.

20. The decoupler of claim 17 wherein the channel estimate matrix comprises a matrix of channel estimates for the plurality of spatial channels between a receiving station and a transmitting station, and  
wherein the receiving station comprises a spatial channel decoupler for  
15 each subcarrier frequency of an orthogonal frequency division multiplexed channel.

21. The decoupler of claim 20 wherein the plurality of spatial channels comprise four spatial channels defined by communication paths between a pair of transmit antennas coupled with the transmitting station and a pair receive  
20 antennas coupled with the receiving station, and  
wherein the channel estimate matrix comprises a 2x2 matrix.

22. A receiver comprising:  
one or more receive antennas to receive a multicarrier signal having components separately transmitted through at least two spatial channels; and  
25 a spatial channel decoupler to separate the separately transmitted signal components of the received multicarrier signal,  
wherein the spatial channel decoupler comprises:

a decomposer to perform a decomposition on a channel estimate matrix to generate a unitary orthogonal matrix and an upper triangular matrix;

a multiplier to multiply received complex signals corresponding with the received multicarrier signals by a conjugate transpose of the unitary orthogonal matrix to generate a z-vector; and

a transmitted signal estimator to estimate transmitted complex signal components of the multicarrier signal from the upper triangular matrix, the upper triangular matrix, and components of the z vector.

23. The receiver of claim 22 wherein the transmitted signal estimator estimates a second transmitted complex signal representing a second signal component of the multicarrier signal from the upper triangular matrix and a second component of the z-vector, and

wherein the transmitted signal estimator further estimates a first transmitted complex signal representing a first signal component of the multicarrier signal based on the estimated transmitted complex signal, the upper triangular matrix, and a first component of the z vector.

24. The receiver of claim 23 wherein the spatial channel decoupler further comprises:

a neighbor calculator to estimate nearest neighbors of the second transmitted complex signal;

a distance calculator to compute Euclidian distances; and

a selector,

wherein the transmitted signal estimator is to re-estimate the first transmitted complex signal for each of the nearest neighbors of the second transmitted complex signal,

wherein the distance calculator computes Euclidian distances between the first transmitted complex signal and the second transmitted complex signal, and between the re-estimated first transmitted complex signals and each neighbor of the second transmitted complex signal, and

wherein the selector selects a pair of signal estimates associated with the lowest computed distance, the selected pair corresponding to a final estimate of the first and second transmitted complex signals.

25. The receiver of claim 22 wherein the received multicarrier signal  
5 comprises two orthogonal frequency division multiplexed signal components of a single orthogonal frequency division multiplexed symbol transmitted substantially simultaneously over the plurality of spatial channels,  
wherein each orthogonal frequency division multiplexed signal component comprises a plurality of symbol-modulated subcarriers transmitted on  
10 orthogonal symbol-modulated subcarriers,  
wherein the spatial channels are non-orthogonal channels, each employing symbol-modulated subcarriers of the same frequencies, and  
wherein the receiver further comprises:  
a spatial channel decoupler associated with each of the subcarriers to  
15 generate transmitted signal estimates for the associated subcarrier.

26. The receiver of claim 25 wherein each of the two orthogonal frequency division multiplexed signal components is to have been modulated with separate data symbols of a single transmitted orthogonal frequency division multiplexed symbol, and  
20 wherein the receiver further comprises:  
a demapper to perform a quadrature amplitude modulation demapping on the first transmitted complex signal and the second transmitted complex signal to generate corresponding first and second bits for each subcarrier; and  
deinterleaver and decoder circuitry to deinterleave and decode the first  
25 and second bits to generate a combined bit stream representing the single transmitted orthogonal frequency division multiplexed symbol.

27. A machine-readable medium that provides instructions, which when executed by one or more processors, cause the processors to perform operations comprising:

- performing a decomposition on a channel estimate matrix to generate a  
5 unitary orthogonal matrix and an upper triangular matrix;
- multiplying received complex signals by a conjugate transpose of the  
unitary orthogonal matrix to generate a z-vector; and
- estimating transmitted complex signal components of the multicarrier  
signal from the upper triangular matrix, the upper triangular matrix, and  
10 components of the z vector.

28. The machine-readable medium of claim 27 wherein the instructions, when further executed by one or more of the processors cause the processors to perform operations further comprising:

- estimating a second transmitted complex signal representing a second  
15 signal component of the multicarrier signal from the upper triangular matrix and  
a second component of the z-vector; and
- estimating a first transmitted complex signal representing a first signal  
component of the multicarrier signal based on the estimated second transmitted  
complex signal, the upper triangular matrix, and a first component of the z  
20 vector.

29. The machine-readable medium of claim 28 wherein the instructions, when further executed by one or more of the processors cause the processors to perform operations, comprising:

- 25 estimating nearest neighbors of the second transmitted complex signal;
- re-estimating the first transmitting complex signal for each of the nearest  
neighbors of the second transmitted complex signal;
- computing Euclidian distances between the first transmitted complex  
signal and the second transmitted complex signal, and between the re-estimated

first transmitted complex signals and each neighbor of the second transmitted complex signal; and

selecting a pair of signal estimates associated with the lowest computed distance, the selected pair corresponding to a final estimate of the first and

5 second transmitted complex signals,

the final estimate of the first and second transmitted complex signals being substantially decoupled from crosstalk between non-orthogonal spatial channels comprising orthogonal symbol modulated orthogonal subcarriers.

30. The machine-readable medium of claim 27 wherein the instructions,  
10 when further executed by one or more of the processors cause the processors to perform operations, wherein the z-vector is equal to a noise factor plus the upper triangular matrix multiplied by an x-vector, and

wherein components of the x-vector represent individual complex signal components transmitted over corresponding individual spatial channels.

15 31. The machine-readable medium of claim 27 wherein the instructions, when further executed by one or more of the processors cause the processors to perform operations, wherein the conjugate transpose of the unitary orthogonal matrix is a hermitian of the unitary orthogonal matrix comprising a hermitian matrix.

20 32. The machine-readable medium of claim 28 wherein the instructions, when further executed by one or more of the processors cause the processors to perform operations, wherein estimating the second transmitted complex signal comprises generating an interference-free estimate of the second transmitted complex signal, the interference-free estimate being substantially free from  
25 contributions of the first transmitted complex signal.

33. The machine-readable medium of claim 27 wherein the instructions, when further executed by one or more of the processors cause the processors to



perform operations, wherein the channel estimate matrix comprises a matrix of channel estimates for the plurality of spatial channels between a receiving station and a transmitting station.

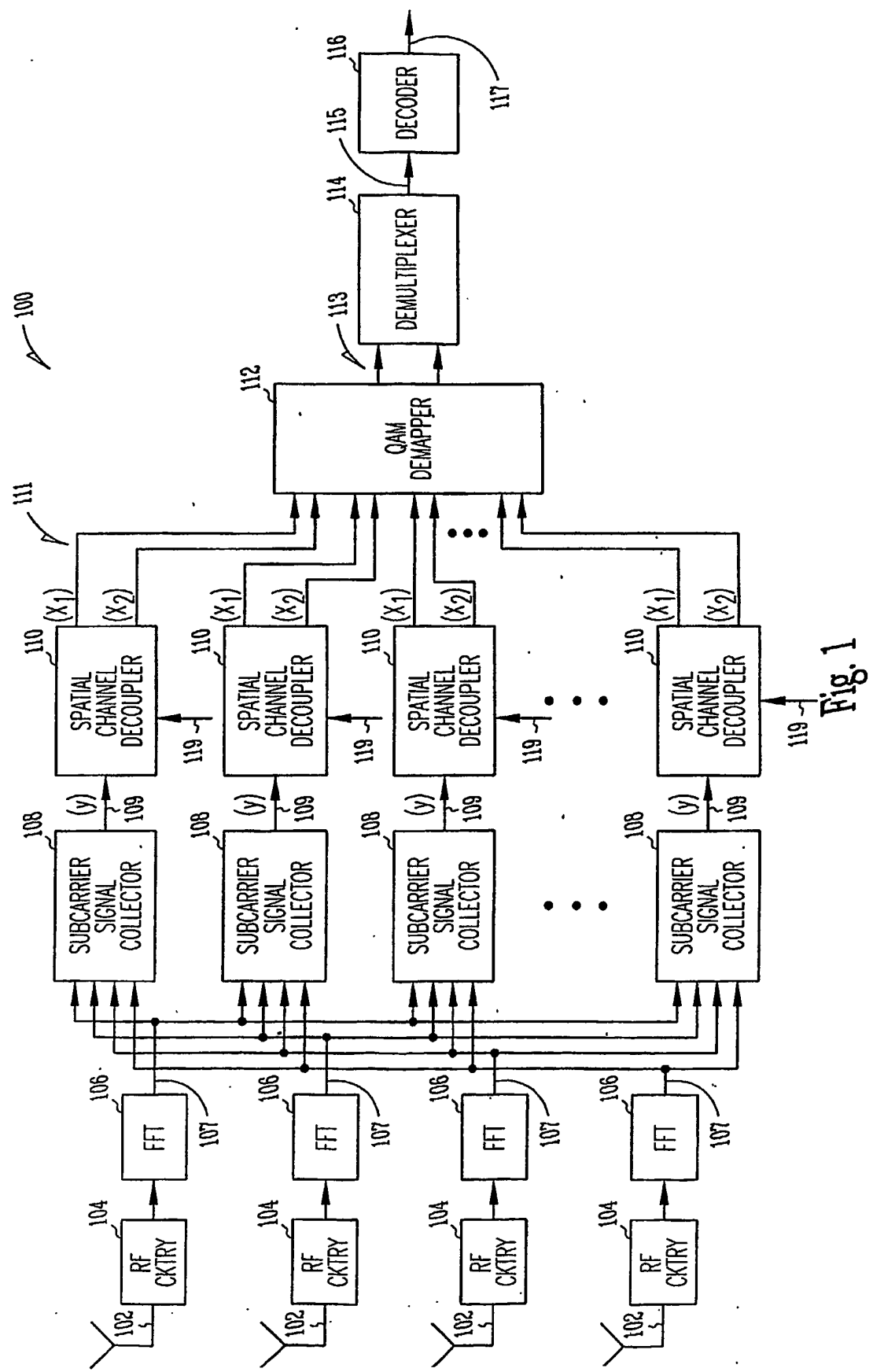


Fig. 1

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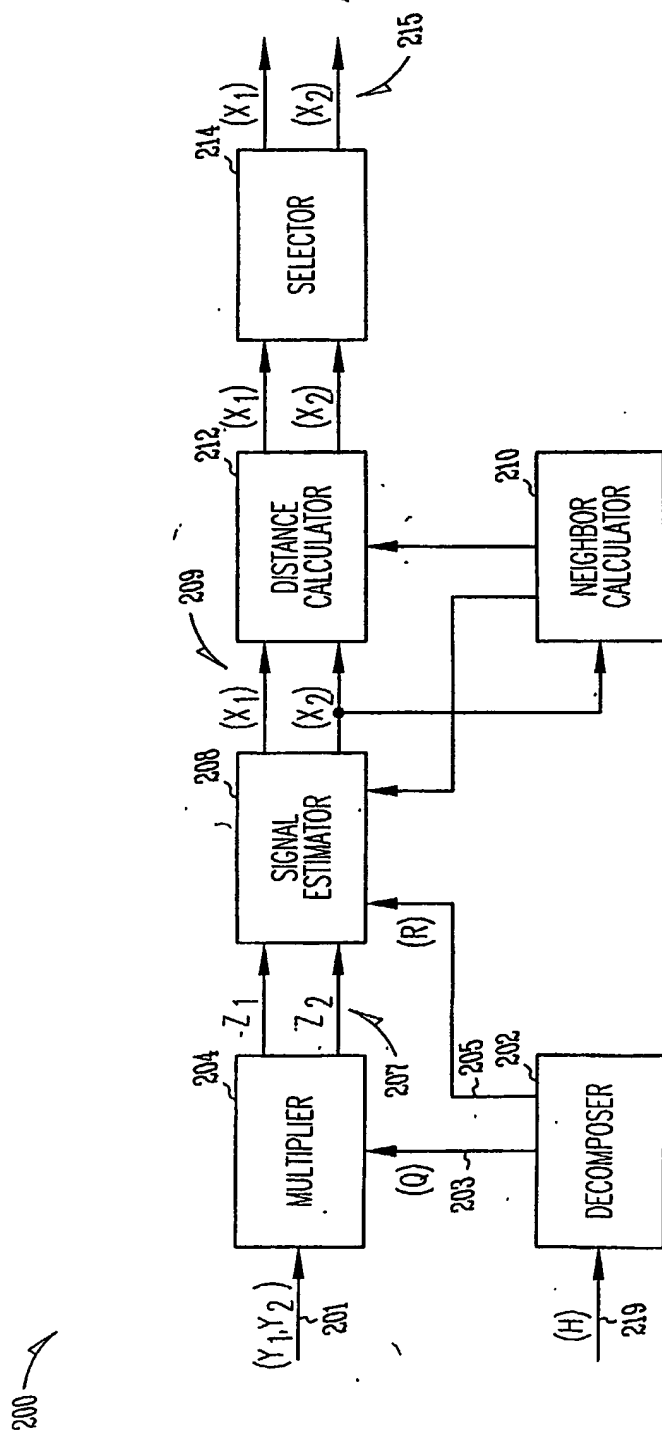


Fig. 2

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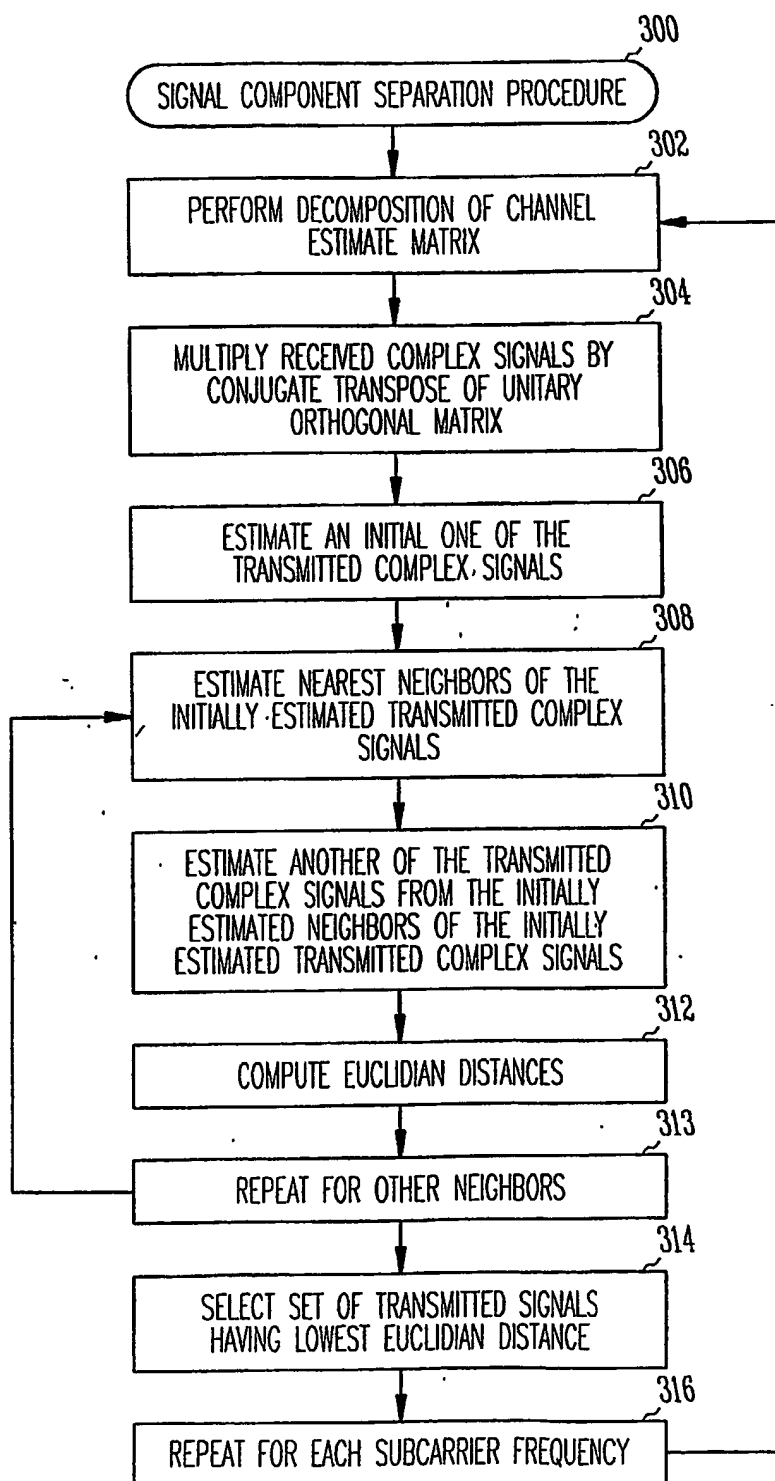


Fig. 3

# INTERNATIONAL SEARCH REPORT

International Application No  
PCT/US2005/001206

## A. CLASSIFICATION OF SUBJECT MATTER

IPC 7 H04L27/26 H04L1/06 H04L25/03

According to International Patent Classification (IPC) or to both national classification and IPC

## B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)

IPC 7 H04L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

EPO-Internal, WPI Data

## C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No
X	<p>BOHNKE R ET AL: "Reduced complexity MMSE detection for BLAST architectures" GLOBECOM'03. 2003 - IEEE GLOBAL TELECOMMUNICATIONS CONFERENCE. CONFERENCE PROCEEDINGS. SAN FRANCISCO, DEC. 1 - 5, 2003, IEEE GLOBAL TELECOMMUNICATIONS CONFERENCE, NEW YORK, NY : IEEE, US, vol. VOL. 7 OF 7, 1 December 2003 (2003-12-01), pages 2258-2262, XP010677757 ISBN: 0-7803-7974-8 the whole document</p> <p style="text-align: center;">----- -/--</p>	<p>1,2,5,6, 14,15, 18,19, 22,23, 27,28, 30-33</p>

☒ Further documents are listed in the continuation of box C

☐ Patent family members are listed in annex

### \* Special categories of cited documents

\*A\* document defining the general state of the art which is not considered to be of particular relevance

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\*O\* document referring to an oral disclosure, use, exhibition or other means

\*P\* document published prior to the international filing date but later than the priority date claimed

\*T\* later document published after the international filing date or priority date and not in conflict with the application but cited to understand the principle or theory underlying the invention

\*X\* document of particular relevance; the claimed invention cannot be considered novel or cannot be considered to involve an inventive step when the document is taken alone

\*Y\* document of particular relevance; the claimed invention cannot be considered to involve an inventive step when the document is combined with one or more other such documents, such combination being obvious to a person skilled in the art.

\*&\* document member of the same patent family

Date of the actual completion of the international search

24 June 2005

Date of mailing of the international search report

05/07/2005

Name and mailing address of the ISA

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## INTERNATIONAL SEARCH REPORT

International Application No

PCT/US2005/001206

## C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
P,X	HIGUCHI K ET AL: "Adaptive selection of surviving symbol replica candidates based on maximum reliability in QRM-MLD for OFCDM MIMO multiplexing" GLOBAL TELECOMMUNICATIONS CONFERENCE, 2004. GLOBECOM '04. IEEE DALLAS, TX, USA 29 NOV.-3 DEC., 2004, PISCATAWAY, NJ, USA, IEEE, vol. 4, 29 November 2004 (2004-11-29), pages 2480-2486, XP010757974 ISBN: 0-7803-8794-5 the whole document	1-33
A	JIANG YUE ET AL: "Channel Estimation and Data Detection for MIMO-OFDM Systems" GLOBECOM'03. 2003 - IEEE GLOBAL TELECOMMUNICATIONS CONFERENCE. CONFERENCE PROCEEDINGS. SAN FRANCISCO, DEC. 1 - 5, 2003, IEEE GLOBAL TELECOMMUNICATIONS CONFERENCE, NEW YORK, NY : IEEE, US, vol. VOL. 7 OF 7, 1 December 2003 (2003-12-01), pages 581-585, XP010677996 ISBN: 0-7803-7974-8 page 581, Abstract - page 583, End of paragraph IV	1-33
A	DAMEN M O ET AL: "On maximum-likelihood detection and the search for the closest lattice point" IEEE TRANSACTIONS ON INFORMATION THEORY, IEEE INC. NEW YORK, US, vol. 49, no. 10, 1 October 2003 (2003-10-01), pages 2389-2402, XP002313805 ISSN: 0018-9448 page 2389, Abstract - page 2397, End of Paragraph V	1-33
A	SEETHALER D ET AL: "Efficient approximate-m1 detection for mimo spatial multiplexing systems by using a 1-D nearest neighbor search" SIGNAL PROCESSING AND INFORMATION TECHNOLOGY, 2003. ISSPIT 2003. PROCEEDINGS OF THE 3RD IEEE INTERNATIONAL SYMPOSIUM ON DARMSTADT, GERMANY 14-17 DEC. 2003, PISCATAWAY, NJ, USA, IEEE, 14 December 2003 (2003-12-14), pages 290-293, XP010729151 ISBN: 0-7803-8292-7 the whole document	3,16,24, 29